

Stability analysis and control design for weakly hard real-time systems How can richer interfaces, e.g. m-k guarantees, be used for the functional verification of control systems

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ESWEEK 2017, October 15, 2017

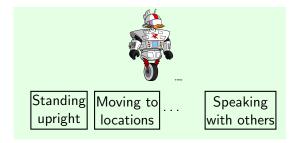


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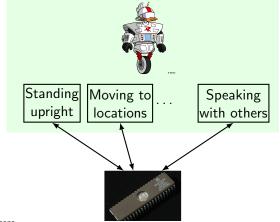
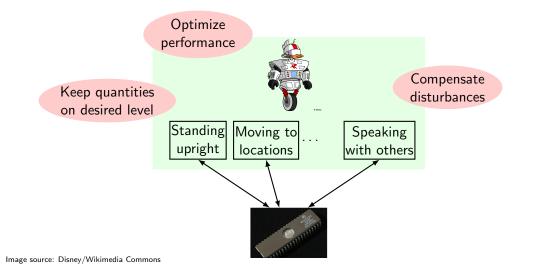
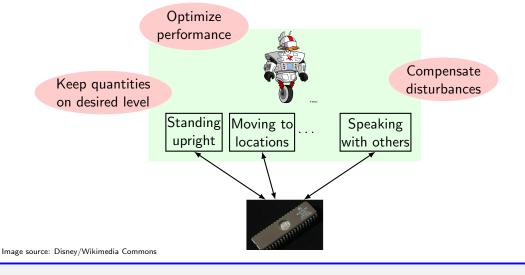


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Is it necessary that all deadlines are met to guarantee properties of the control system?



- 1 Functional verification of control systems
- 2 How to use weakly hard real-time constraints in control systems
- 3 Stability analysis with weakly hard real-time constraints
- 4 Controller design with weakly hard real-time constraints
- **5** Summary



1 Functional verification of control systems

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6 Summary

Functional verification - Stability 🔷



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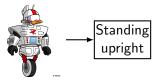


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Functional verification - Stability

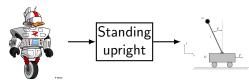


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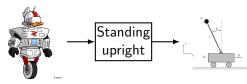


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This can be used to motivate/illustrate a major property of control systems: (Asymptotic) stability

Functional verification - Stability 🔷

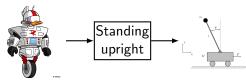
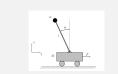


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This can be used to motivate/illustrate a major property of control systems: (Asymptotic) stability

The upright equilibrium is unstable! \Rightarrow One needs a controller to stabilize it.



Goal: <u>Guarantee</u> asymptotic stability of the upright equilibrium

Functional verification - Stability

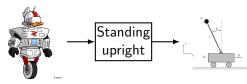


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This can be used to motivate/illustrate a major property of control systems: (Asymptotic) stability

The upright equilibrium is unstable! \Rightarrow One needs a controller to stabilize it.



Goal: <u>Guarantee</u> asymptotic stability of the upright equilibrium *with richer interfaces*.

Functional verification - Stability

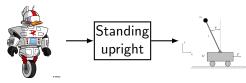


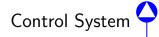
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Asymptotic Stability

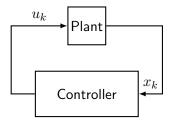
The equilibrium at the origin of $x_{k+1} = f(x_k)$ is asymptotically stable if it is stable and if δ can be chosen such that $||x_{k_0}|| < \delta$ implies that $||x_k|| \to 0$ when $k \to \infty$.



This talk: Linear time-invariant systems with full state available for feedback.

State-space representation of the plant:

$$x_{k+1} = Ax_k + Bu_k$$





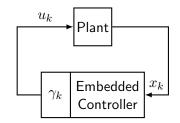
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$$u_k = f(\gamma_k, x_k, (u_{k-1})), \gamma_k \in \{0, 1\}$$



Stabilization

Can we derive a (static state feedback) controller that renders the origin of the closed loop asymptotically stable?

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Performance

Can we derive a controller that minimizes a given cost functional?

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Performance

Can we derive a controller that minimizes a given cost functional?

Robustness

Can we derive a controller that is robust to uncertainties and disturbances?

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Up to now:

• Stability and Stabilization as important properties and tasks for functional verification of control systems.

Open tasks:

- Show that we can include richer interfaces in control system models.
- Analyze stability with richer interfaces.
- Design controllers with richer interfaces.



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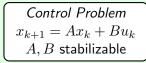
6 Summary

Motivation for using richer interfaces \bigcirc

Control Problem $x_{k+1} = Ax_k + Bu_k$ A, B stabilizable

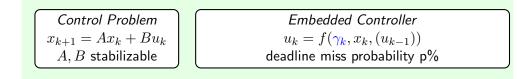
Control Problem $x_{k+1} = Ax_k + Bu_k$ A, B stabilizable

Embedded Controller $u_k = f(\gamma_k, x_k, (u_{k-1}))$ model of the process γ_k

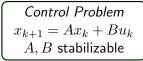


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- Window based description \Rightarrow promising idea to bridge the gap

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- Window based description \Rightarrow promising idea to bridge the gap

Goal

Include window based descriptions in the model of embedded control systems.

Characterization of binary sequences γ by satisfaction of constraints λ " $(\gamma \vdash \lambda)$ ":

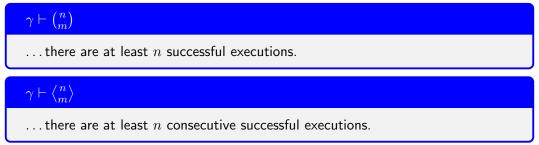
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In any window of m consecutive executions, \ldots



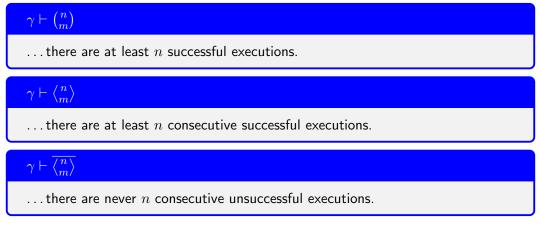
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Partial order of constraints

Satisfaction Set

The satisfaction set of length N of a constraint λ , denoted $\mathcal{S}^N(\lambda)$, is the set of all sequences $\alpha \in \{0,1\}^N$ that satisfy λ . Formally,

$$\mathcal{S}^{N}(\lambda) := \left\{ \alpha \in \{0,1\}^{N} : \alpha \vdash \lambda \right\}.$$

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Partial Order

Given two constraints λ', λ . We say that λ' is harder than λ (λ is weaker than λ'), denoted $\lambda' \leq \lambda$ if all sequences that satisfy λ' also satisfy λ . Formally,

 $\lambda' \preceq \lambda \Leftrightarrow \mathcal{S}^{\infty}(\lambda') \subseteq \mathcal{S}^{\infty}(\lambda).$

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Example:

At least 3 succ. executions in a row is harder than at least 2 succ. executions in a row.

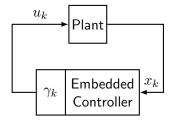
Model of the closed loop with richer interfaces

State-space representation of the plant:

 $x_{k+1} = Ax_k + Bu_k$

Representation of the embedded controller:

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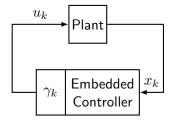
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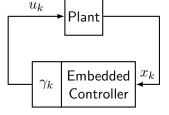
Actuation strategies:

zero strategy

$$u_k = \gamma_k K x_k$$

hold strategy

$$u_k = \gamma_k K x_k + (1 - \gamma_k) u_{k-1}$$



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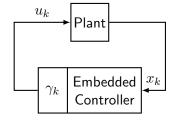
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Up to now:

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- We can include WHRT constraints in control system models.

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Probabilistic description of γ :

Stabilizable in the mean square sense if and only if the success probability is greater than $1 - \frac{1}{a^2}$.



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- Then x(k+1) = 0 after the first successful execution.



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The system is asymptotically stable for all sequences $\gamma \vdash \lambda \preceq {\binom{1}{m}}$, where m might be arbitrarily large but finite.

Closed loop as switched system \bigcirc

The closed loop can be modeled as a linear discrete-time switched system,

$$\xi_{k+1} = \mathcal{A}_{\gamma_k} \xi_k, \quad \xi_k \in \mathbb{R}^{n_{\xi}}, \gamma_k \in \{0, 1\}, \gamma \vdash \lambda$$

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Zero control:

- $u_k = \gamma_k K x_k$
- $\xi_k = x_k \to n_\xi = n$
- $\mathcal{A}_0 = A$
- $\mathcal{A}_1 = A + BK$

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$$u_k = \gamma_k K x_k + (1 - \gamma_k) u_{k-1}$$

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$$\xi_k = [x_k, u_{k-1}] \rightarrow n_{\xi} = n + n_u$$

• $\mathcal{A}_0 = \begin{bmatrix} A & B \\ 0 & I \end{bmatrix}, \mathcal{A}_1 = \begin{bmatrix} A + BK & 0 \\ K & 0 \end{bmatrix}$

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Notation

Given a sequence $w \in \{0,1\}^m$, the state evolves as $x_{k+|w|} = \prod_{i=|w|}^1 \mathcal{A}_{w_i} x_k$. Hence, we use $\mathcal{A}^w := \prod_{i=|w|}^1 \mathcal{A}_{w_i}$. Example: $\mathcal{A}^{1100101} = \mathcal{A}_1 \mathcal{A}_0 \mathcal{A}_1 \mathcal{A}_0 \mathcal{A}_0 \mathcal{A}_1 \mathcal{A}_1$.

Theorem 1 and procedure for stability analysis

Theorem 1

Given the constraint λ and the matrices $\mathcal{A}_0, \mathcal{A}_1 \in \mathbb{R}^{n_{\xi} \times n_{\xi}}$. If there exists a symmetric, positive definite matrix $\mathcal{P} \in \mathbb{R}^{n_{\xi} \times n_{\xi}}$ such that

$$(\mathcal{A}^w)^{\mathsf{T}} \mathcal{P} \mathcal{A}^w - \mathcal{P} < 0, \forall w \in \mathcal{S}^m(\lambda),$$

then the closed loop is asymptotically stable for all sequences $\gamma \vdash \lambda' \preceq \lambda$.

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Procedure to check whether the origin of the closed-loop is asy. stable:

- Compute $\mathcal{A}_0, \mathcal{A}_1$ according to the actuation strategy.
- Derive all sequences in $\mathcal{S}^m(\lambda)$.
- Compute all \mathcal{A}^w for $w \in \mathcal{S}^m(\lambda)$.
- Find a positive definite matrix ${\cal P}$ such that all LMI's are satisfied.

Theorem 2 and procedure for stability analysis

Theorem 2

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- Derive all sequences in $S^m(\lambda)$ and all combinations $w_1w_2 \vdash \lambda$.
- Compute all \mathcal{A}^w for $w \in \mathcal{S}^m(\lambda)$.
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Theorem 2

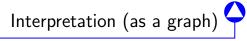
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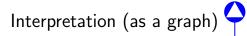
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- Find |S^m(λ)| positive definite matrices P_w such that LMI's are satisfied (combinations s.t. w₁w₂ ⊢ λ). ⇒ increased possibility to find a solution



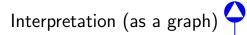
Interpretation:

• Theorem 2 takes into account that the concatenation of sequences is actually constrained.



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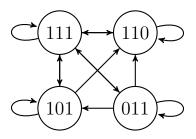
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- This can be represented by a graph.



Interpretation:

- Theorem 2 takes into account that the concatenation of sequences is actually constrained.
- This can be represented by a graph.

Example: $\lambda = \begin{pmatrix} 2 \\ 3 \end{pmatrix}$





$$x_{k+1} = \begin{bmatrix} 0 & 1\\ 1 & 1 \end{bmatrix} x_k + \begin{bmatrix} 1\\ 1 \end{bmatrix} u_k$$

Embedded Controller:

$$K = egin{bmatrix} -0.85 & -0.85 \end{bmatrix}$$
 used for zero and hold control



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Embedded Controller:

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Goal

Check stability for
$$\binom{1}{2}$$
, $\binom{1}{3}$, $\binom{1}{4}$, and $\binom{2}{3}$.



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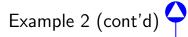
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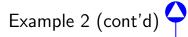
Remark:
$$\binom{2}{3} \preceq \binom{1}{2} \preceq \binom{1}{3} \preceq \binom{1}{4}$$
.



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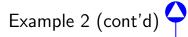
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|---|--|-----------------------|--|
| | Zero strategy | Hold strategy | |
| $\lambda = \begin{pmatrix} 2\\ 3 \end{pmatrix}$ $\lambda = \begin{pmatrix} 1\\ 2 \end{pmatrix}$ $\lambda = \begin{pmatrix} 1\\ 2 \end{pmatrix}$ | Theorem 1√ | Theorem $1\checkmark$ | |
| $\lambda = \begin{pmatrix} 1 \\ 3 \end{pmatrix}$ $\lambda = \begin{pmatrix} 1 \\ 4 \end{pmatrix}$ | | | |



$$x_{k+1} = \begin{bmatrix} 0 & 1 \\ 1 & 1 \end{bmatrix} x_k + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u_k$$

Embedded Controller:

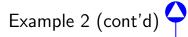
| | $K = egin{bmatrix} -0.35 & -0.85 \end{bmatrix}$ used for zero and hold control | | |
|---|--|-----------------------|--|
| | Zero strategy | Hold strategy | |
| $\lambda = \begin{pmatrix} 2 \\ 3 \end{pmatrix}$ | Theorem $1\checkmark$ | Theorem $1\checkmark$ | |
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| $\lambda = \begin{pmatrix} 2\\ 3 \end{pmatrix}$ $\lambda = \begin{pmatrix} 1\\ 2 \end{pmatrix}$ | Theorem $1\checkmark$ | Theorem 2√ | |
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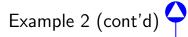


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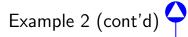


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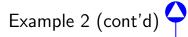
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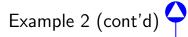


λ λ λ

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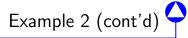


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$$K = \begin{bmatrix} -0.35 & -0.85 \end{bmatrix}$$
 used for zero and hold control

Up to now:

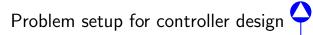
- Stability and Stabilization as important properties and tasks for functional verification of control systems.
- We can include WHRT constraints in control system models.
- We can analyze stability with WHRT constraints.

Open tasks:

• Design controllers with richer interfaces.



- 1 Functional verification of control systems
- 2 How to use weakly hard real-time constraints in control systems
- 3 Stability analysis with weakly hard real-time constraints
- 4 Controller design with weakly hard real-time constraints
- **6** Summary



Up to now we considered all components as given and checked whether the origin of the closed loop is asymptotically stable.

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In the remainder:

Static state feedback stabilization

Given

- a weakly hard real-time constraint λ ,
- a discrete-time linear control system (A, B stabilizable),
- and an actuation strategy (zero resp. hold),

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In the remainder:

Static state feedback stabilization

Given

- a weakly hard real-time constraint λ ,
- a discrete-time linear control system (A, B stabilizable),
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find K s.t. the origin of the closed loop system with $u_k = \gamma_k K x_k$ resp. $u_k = \gamma_k K x_k + (1 - \gamma_k) u_{k-1}$ is asymptotically stable for all $\gamma \vdash \lambda$.





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- For the computation of a controller one needs to take care that the dimension of the controller does not vary with the chosen strategy, i.e., dim K may not be dependent on dim ξ.
- \Rightarrow This is covered by the introduction of an alternative discretization

New sequences are derived based on $\boldsymbol{\gamma}$:

- $(\tau_k)_{k\in\mathbb{N}}$ contains all time instants when the computation is executed in time.
- $(\alpha_k)_{k\in\mathbb{N}}$ contains the number of deadline misses after an executed computation.

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• Example:

$$\gamma = (0, 1, 0, 0, 1, 0, 1, \dots)$$

 $\tau = (1, 4, 6, \dots)$
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$$\tilde{\xi}_{k+1} = \tilde{\mathcal{A}}_{\alpha_k} \tilde{\xi}_k$$

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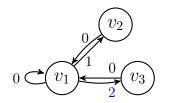
with $\tilde{\xi}_k := \xi_{\tau_k}$.

- Stability of the original closed loop system can be deduced from the \tilde{x} dynamics of the restricted system.
- The \tilde{x} dynamics are in the form $A_l^x + B_l^x K$.



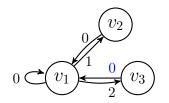






$$\gamma = (1, 0, 0, 1,$$
$$\alpha = (2,$$



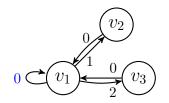


$$\gamma = (1, 0, 0, 1, 1, 1)$$

 $\alpha = (2, 0, 1)$

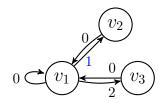


Graph representation (Example $\gamma \vdash {\binom{2}{5}}$):



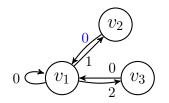


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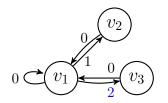
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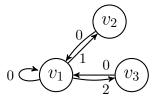


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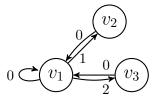
Recall previous example:

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Graph representation (Example $\gamma \vdash {\binom{2}{5}}$):



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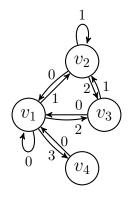
$$\gamma = (0, 1, 0, 0, 1, 0, 1, ...)$$

 $\tau = (1, 4, 6, ...)$
 $\alpha = (2, 1, ...)$

This sequence does not satisfy the constraint!

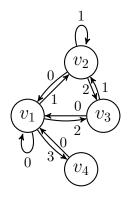
Examples for other constraints

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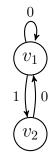


Examples for other constraints

Example $\gamma \vdash \binom{2}{5}$:



Example
$$\gamma \vdash \overline{\langle 2 \atop 5 \rangle}$$
:





Theorem 3

isto

Given a WHRT constraint λ and its graph representation \mathcal{G} ,



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Theorem 3 (Stabilization)

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Given a WHRT constraint λ and its graph representation \mathcal{G} , the discrete-time linear control system is state feedback stabilizable under the zero (resp. hold) strategy w.r.t. all $\lambda' \leq \lambda$, if there exist n_V symmetric matrices $S_1, \ldots, S_{n_V} \in \mathbb{R}^{n \times n}$ and matrices $G \in \mathbb{R}^{n \times n}$, $F \in \mathbb{R}^{n_u \times n}$, satisfying

$$\begin{bmatrix} G+G^{\mathsf{T}}-S_i & (A_l^xG+B_l^xF)^{\mathsf{T}} \\ A_l^xG+B_l^xF & S_j \end{bmatrix} > 0, \; \forall (i,j,l) \in E.$$

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A stabilizing state feedback controller is given by $K = FG^{-1}$.





1 Derive a labeled graph \mathcal{G} for the constraint λ .



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- **2** Compute the matrices A_l, B_l for all labels,

$$\begin{split} A_l &= A^{l+1} \\ B_l &= \begin{cases} A^l B & \text{zero strategy} \\ \sum_{i=0}^l A^i B & \text{hold strategy.} \end{cases} \end{split}$$



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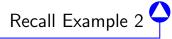
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4 Compute a stabilizing state feedback controller for all $\lambda' \leq \lambda$,

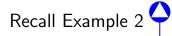
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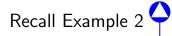
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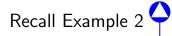
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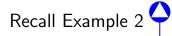
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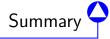
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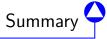
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| $\lambda = \begin{pmatrix} 1 \\ 4 \end{pmatrix}$ | no stabilization with Thm 4🗡 | stabilization with $K = \begin{bmatrix} -0.29 & -0.48 \end{bmatrix}$ 🗸 |

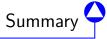


- 1 Functional verification of control systems
- 2 How to use weakly hard real-time constraints in control systems
- 3 Stability analysis with weakly hard real-time constraints
- ④ Controller design with weakly hard real-time constraints
- **5** Summary

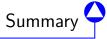




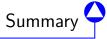
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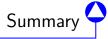
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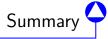
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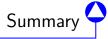
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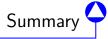


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Thank you for your attention!



Stability analysis and control design for weakly hard real-time systems How can richer interfaces, e.g. m-k guarantees, be used for the functional verification of control systems

Steffen Linsenmayer (partially work of and with R. Blind and F. Allgöwer)

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