

# On deciding stability of high frequency amplifiers

L. Baratchart, S. Fueyo, G. Lebeau, J.-B. Pomet

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*Journées McTAO-CAGE*

The logo for Inria, consisting of the word "Inria" written in a stylized, red, cursive script.

# Summary

- 1 Stability Conditions for Time-varying Networks of Telegrapher's Equations or Difference Delay Equations
- 2 Stability of high frequency amplifiers

## Example time-varying Networks of Telegrapher's equations

Telegrapher's equations :

$$\begin{cases} C_k \frac{\partial v_k(t, x)}{\partial t} = -\frac{\partial i_k(t, x)}{\partial x}, \\ L_k \frac{\partial i_k(t, x)}{\partial t} = -\frac{\partial v_k(t, x)}{\partial x}, \end{cases} \quad (t, x) \in \Omega,$$

$$\Omega = \{(t, x) \in \mathbb{R}^2, 0 < x < 1 \text{ and } 0 < t < +\infty\},$$

With  $k \in \{1, \dots, N\}$  where  $N$  positive integer denoting the number of Telegrapher's equations.

# General time-varying Networks of Telegrapher's equations

$$v(x, t) = \begin{pmatrix} v_1(x, t) \\ \vdots \\ v_N(x, t) \end{pmatrix}, \quad i(x, t) = \begin{pmatrix} i_1(x, t) \\ \vdots \\ i_N(x, t) \end{pmatrix}$$

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$$\mathbf{V}(t) = \begin{pmatrix} v(t, 0) \\ \vdots \\ v(t, 1) \end{pmatrix}, \quad \mathbf{I}(t) = \begin{pmatrix} -i(t, 0) \\ \vdots \\ i(t, 1) \end{pmatrix},$$

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## Assumption (Dissipativity)

The map  $t \mapsto \mathbf{A}(t)$  is continuous and bounded.

$$\mathbf{A}(t) + \mathbf{A}^*(t) \geq \alpha Id, \quad \alpha > 0 \quad t \in \mathbb{R}.$$

## Equivalence with linear time-varying delay system

$$\bullet \begin{pmatrix} x_1(t) \\ \vdots \\ x_N(t) \\ y_1(t) \\ \vdots \\ y_N(t) \end{pmatrix} = (I + \mathbf{A}(t) \mathbf{K})^{-1} (I - \mathbf{A}(t) \mathbf{K}) P_2 \begin{pmatrix} x_1(t - \tau_1) \\ \vdots \\ x_N(t - \tau_N) \\ y_1(t - \tau_1) \\ \vdots \\ y_N(t - \tau_N) \end{pmatrix} .$$

- $\mathbf{K}$  diagonal positive matrix and  $P_2$  permutation matrix.
- One to one linear relation between  $i_k, v_k$  and  $x_k, y_k$ .

# Generalities on the linear time-varying delay systems

$$z(t) = \sum_{i=1}^N D_i(t) z(t - \eta_i) \quad t \geq 0. \quad (1)$$

## Theorem

Let  $\phi$  be an element of  $L^2([-\eta_N, 0], \mathbb{R}^d)$ .

(i) There is a unique solution  $z$  to (1) in  $L^2_{loc}([-\eta_N, +\infty), \mathbb{R}^d)$  meeting the initial condition  $z|_{[-\eta_N, 0]} = \phi$ .

(ii) Moreover if  $\phi$  is continuous on  $[-\eta_N, 0]$  with

$$\phi(0) = \sum_{i=1}^N D_i(0) \phi(-\eta_i), \text{ then } z \in C^0([-\eta_N, +\infty), \mathbb{R}^d).$$



## Definition Stability

$$z(t) = \sum_{i=1}^N D_i(t) z(t - \eta_i) \quad (2)$$

## Definition

The system (2) is  $L^2$  (resp.  $C^0$ ) exponentially stable if there exists  $\gamma, K > 0$  such that :

$$\|z(t + \cdot)\|_{L^2([- \eta_N, 0], \mathbb{R}^d)} \leq Ke^{-\gamma t} \|z(\cdot)\|_{L^2([- \eta_N, 0], \mathbb{R}^d)}, \quad t \geq 0$$

(resp.  $\|z(t + \cdot)\|_{C^0([- \eta_N, 0], \mathbb{R}^d)} \leq Ke^{-\gamma t} \|z(\cdot)\|_{C^0([- \eta_N, 0], \mathbb{R}^d)}, \quad t \geq 0$ ).

## Stability linear **time-invariant** delay system

$$z(t) = \sum_{i=1}^N D_i z(t - \eta_i) \quad (3)$$

Theorem (Henry-Hale Theorem, [Hen74, HVL93])

*The following properties are equivalent :*

- 1 System (3) is  $L^2$  exponentially stable.
- 2 System (3) is  $C^0$  exponentially stable.
- 3 There exists  $\beta < 0$  for which

$Id - \sum_{i=1}^N D_i e^{-\lambda \eta_i}$  is invertible for all  $\lambda \in \mathbb{C}$  such that  $\Re(\lambda) > \beta$ .

# Stability **time-invariant** network of Telegrapher's equation

## Theorem

*Under the dissipativity assumption, the delay difference system associated to the network of Telegrapher's equations with time-invariant boundaries is  $C^0$  exponentially stable.*

*Proof :*

①

$$\sum_{i=1}^N D_i e^{-\lambda\tau_i} = [Id + \mathbf{AK}]^{-1} [Id - \mathbf{AK}] P_2 \begin{bmatrix} e^{-\lambda\eta_1} & & & \\ & \ddots & & \\ & & \ddots & \\ & & & e^{-\lambda\eta_N} \end{bmatrix} .$$

② There exists  $\|\cdot\|_{\mathbf{K}}$ ,  $\left\| \sum_{i=1}^N D_i e^{-\lambda\tau_i} \right\|_{\mathbf{K}} \leq e^{-\Re(\lambda)\eta_1}$  .

# Time-varying linear delay system?

The Henry-Hale theorem is no longer true for time-varying linear difference delay system.

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Few references :

- Stability criteria through Perron-Froebenius theorem  
Ngoc and Huy [NH15]  
→ Not suitable for our case
- Stability criteria through joint spectral radius  
Chitour, Mazanti and Sigalotti [CMS16]  
→ Combinatorics too complicated

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Our Strategy :

- Prove  $L^2$  stability through Lyapunov functional for system of PDEs
- Prove that stability  $L^2$  is equivalent to the stability  $C^0$  for difference delay system ([CMS16] or [BFLP19])
- Conclude that we have the  $C^0$  stability for our system of PDEs.

## L2 stability for PDE

## Theorem

*Under the dissipativity, the delay system associated to the time-varying network of telegrapher's equation is  $L^2$  exponentially stable.*

- 1 Lyapunov function :

$$E_k(t) = \frac{1}{2} \int_0^1 [C_k v_k^2(t, x) + L_k i_k^2(t, x)] dx, \quad E(t) = \sum_{k=1}^N E_k(t).$$

- 2  $\frac{d}{dt} E(t) \leq -\alpha \sum_{k=1}^N [i_k^2(t, 0) + i_k^2(t, 1)].$

- 3 We can prove :  $E(t) \leq Ke^{-\gamma t} E(0), \gamma, K > 0.$



# Equivalence $C^0$ and $L^2$ stability for difference delay system

Proposition ([CMS16], BFLP 2019)

*A periodic delay system is  $L^2$  exponentially stable if and only if it is  $C^0$  exponentially stable.*

*Sketch of proof :*

①

$$z(t) = \sum_{q=1}^{+\infty} M_q(t) \phi(t - \sigma_q), \quad t \geq 0,$$

for at most  $Ct^N$  non zero terms.

② Using suitable test initial data,  $L^2$  or  $C^0$  stability implies :

$$\|M_q(s)\| \leq C_0 e^{-\gamma s}, \quad \gamma, C_0 > 0.$$

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# Stability for a subclass of delay system

$$z(t) = \sum_{i=1}^N D_i(t) z(t - \eta_i), \eta_i \text{ distinct}$$

## Theorem

If we have :

- 1 Disjoint column properties,
- 2 The sum of the matrices  $D_i(t)$  is uniformly contractive:

$$\left\| \sum_{i=1}^N D_i(t) \right\|_2 \leq \nu, \nu < 1,$$

then the difference delay system is  $C^0$  stable.

# Summary

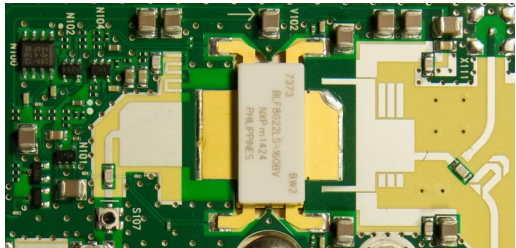
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# Inside

An amplifier is made of interconnected

- 1 resistors, inductors, capacitors,
- 2 diodes/transistors,
- 3 lossless transmission lines which cannot be neglected at high frequency inducing delays.

Forcing periodic signal ►► periodic solution in the amplifier ►► amplified signal.



# Motivation

- Amplifiers at high frequency are ubiquitous (Cell phones, relays...). They need to be quick to design and produced in large quantities.
- Computer-assisted design (CAD) and simulation before production.
- Powerful “frequency simulation” tools give a reliable prediction of the response, but that response might be unstable.
- **Need** for a tool to predict stability/unstability in the frequency domain.



# Harmonic Balance

The Harmonic Balance method, through Fourier development, Laplace transform and fixed point methods permits to :

- approximate the periodic solution of the circuit,
- linearize the circuit around the periodic solution,
- give a frequency response to a periodic signal wich disturbs the linearized circuit.

Harmonic balance method : Numerically implemented.

Our focus : Structure of the harmonic transfer function, its singularities, links with stability.

General system,  $T$ -periodic :

$$\left\{ \begin{array}{l} \frac{dx(t)}{dt} = A_1(t)x(t) + \sum_{i=0}^N B_{1,i}(t)y(t - \tau_i) \\ y(t) = \sum_{i=1}^N B_{2,i}(t)y(t - \tau_i) + A_2(t)x(t), \quad t \geq s, \end{array} \right.$$

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- $L^2 := \mathbb{R}^n \times L^2([- \tau_N, 0], \mathbb{R}^k)$ .

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$$\left. \begin{array}{l} L^2 \text{ exponential} \\ \text{stability} \end{array} \right\} \Leftrightarrow \left\{ \begin{array}{l} Sp(U(T, 0)) \text{ included in} \\ \text{disc of radius } r < 1 \end{array} \right.$$

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- Monodromy operator  $U(T, 0)$
- $L^2$  exponential stability  $\} \Leftrightarrow \left\{ \begin{array}{l} Sp(U(T, 0)) \text{ included in} \\ \text{disc of radius } r < 1 \end{array} \right.$

## Behaviour at high frequency

High frequency system :

$$\begin{cases} x(t) = 0 \\ y(t) = \sum_{i=1}^N B_{2,i}(t)y(t - \tau_i), \quad t \geq s, \end{cases}$$

- $\tilde{L}^2 := \{0_n\} \times L^2([-\tau_N, 0], \mathbb{R}^k)$ .
- Solution operator  $V(t, s) : \tilde{L}^2 \rightarrow \tilde{L}^2$ .
- Monodromy operator  $V(T, 0)$ .
- 

$$\left. \begin{array}{l} L^2 \text{ or } C^0 \text{ exponential} \\ \text{stability} \end{array} \right\} \Leftrightarrow \{ \text{High Frequency dissipativity} \}$$



## Compact perturbation

### Lemma

*We have :*

$$U(t, s) = V(t, s)P + K(t, s), \quad t \geq s$$

*with  $K(t, s)$  compact operator  $L^2 \rightarrow L^2$  for all  $t, s$  and  $P$  the canonical projection  $L^2 \rightarrow \tilde{L}^2$ .*

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### Theorem

*The monodromy operator  $U(T, 0)$  possesses at most a finite number of eigenvalues  $\zeta_1, \dots, \zeta_n$  outside a disk of a radius strictly less than 1.*

# $L^2$ stability equivalent to $C^0$ stability

## Theorem

*The general system is  $L^2$  exponentially stable if and only if it is  $C^0$  exponentially stable.*

The stability of high frequency amplifiers depends on a finite number of unstable eigenvalues.

## Input-Output system

$$\left\{ \begin{array}{l} \frac{dx(t)}{dt} = A_1(t)x(t) + \sum_{i=0}^N B_i^1(t)y(t - \tau_i) + C_1(t)u(t) \\ y(t) = \sum_{i=1}^N B_i^2(t)y(t - \tau_i) + A_2(t)x(t) + C_2(t)u(t) \\ z(t) = \sum_{i=1}^N B_i^3(t)y(t - \tau_i) + A_3(t)x(t) + C_3(t)u(t), \quad t \geq 0, \end{array} \right.$$

- $x(t), y(t), z(t) = 0$  for  $t < 0$ ,
- Input  $u \in L^2_{loc}([0, +\infty), \mathbb{R})$  current perturbation, output  $z$  the voltage,
- All coefficients are  $T$  - periodic.

# Harmonic Transfer Function

- $X(t, \alpha)$  response at time  $t$  to an impulse at time  $\alpha$   
$$z(t) = \int_0^t X(t, \alpha)u(\alpha)d\alpha$$

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## Definition (Harmonic Transfer Function HTF)

The infinite matrix  $H(s)$  defined by  $H_{m,n}(s) := G_{n-m}(s + \frac{2i\pi m}{T})$  for  $s \in \mathbb{C}$  is called the harmonic transfer function.



# Harmonic Transfer Function

$$Z(s) := \int_0^{+\infty} z(t)e^{-st} dt \text{ and } U(s) := \int_0^{+\infty} u(t)e^{-st} dt.$$

$$\begin{pmatrix} \vdots \\ Z(s + i\omega_0) \\ Z(s) \\ Z(s - i\omega_0) \\ \vdots \end{pmatrix} = \begin{pmatrix} \ddots & \vdots & \vdots & \vdots & \ddots \\ \cdots & G_2(s - i\omega_0) & G_1(s) & G_0(s + i\omega_0) & \cdots \\ \cdots & G_1(s - i\omega_0) & G_0(s) & G_{-1}(s + i\omega_0) & \cdots \\ \cdots & G_0(s - i\omega_0) & G_{-1}(s) & G_{-2}(s + i\omega_0) & \cdots \\ \ddots & \vdots & \vdots & \vdots & \ddots \end{pmatrix} \begin{pmatrix} \vdots \\ U(s + i\omega_0) \\ U(s) \\ U(s - i\omega_0) \\ \vdots \end{pmatrix}$$

- HTF is an operator valued analytic map  
(values: continuous ops  $l^2(\mathbb{Z}) \rightarrow l^2(\mathbb{Z})$ )
- its entries  $\{G_n\}$  are complex valued analytic maps

# Structure of the Harmonic Transfer Function

Define  $z_{j,k} = \frac{\ln(\zeta_j) + 2ik\pi}{T}$  for  $j$  in  $\{1\dots n\}$ ,  $k$  in  $\mathbb{Z}$ .

## Theorem

In  $\{s \in \mathbb{C}, \Re(s) \geq \gamma\}$  for some  $\gamma < 0$ ,

- $H$  is a meromorphic operator  $l^2(\mathbb{Z}) \rightarrow l^2(\mathbb{Z})$  with possibly poles at  $\{z_{j,k}, j \in \{1\dots n\}, k \in \mathbb{Z}\}$ .

Under observability/controllability assumptions,

- for all  $j$ , there is at least a  $k$  such that  $z_{j,k}$  is a pole of  $H$ , and also a pole of one  $G_n$ .






If no **pole** in right half plane, exponential  $C^0$  stability.

## Contribution

- Advances in stability of time-varying delay systems coming from a network of Telegrapher's equations
- Math. foundation of HF amplifiers stability decision in CAD
- Projection on the unstable part and rational approximation to find the poles

## Open questions

- Generalization of the Henry-Hale to the periodic difference delay system?
- For fixed  $j$ , which  $z_{j,k}$  is a pole of which  $G_n$ ?  
(In practice, few  $G_n$  are computed.)
- Bound on the number of unstable poles?
- May the (stable) singularities of the  $G_n$ 's be other than poles ?

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